

# Reviews on Inductor, AC-Circuits, Electromagnetic Waves, and Visible light

## SCPY152, Second Semester 2021-22

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# Topics

1. Inductor
2. Magnetic field energy
3. Complex impedance and AC circuits
4. RLC resonance
5. Electromagnetic waves
6. Visible light

## Inductor

- ▶ Consider a solenoid of  $N$  turn, length  $L$  (so that  $n = N/L$ ) and cross section area  $A$ . When a current  $i(t)$  pass through the wire the magnetic field produced inside  $B(t) = \mu_0 n i(t)$  and magnetic flux through the solenoid is

$$\Phi_B(t) = B(t)A = \mu_0 n A i(t)$$

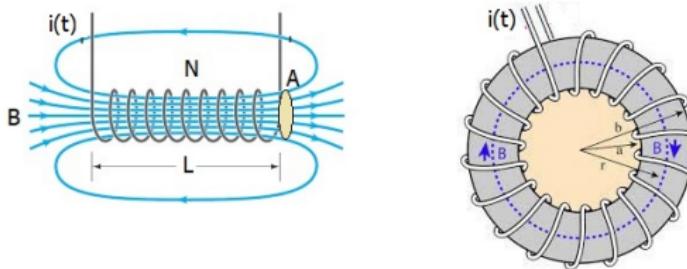
The change in time of this flux induces emf inside the wire as

$$\mathcal{E} = -N \frac{d\Phi_B(t)}{dt} = -\mu_0 n N A \frac{di(t)}{dt} \equiv -V_L(t) \quad (1)$$

$$V_L(t) = \mu_0 n N A \frac{di(t)}{dt} = L \frac{di(t)}{dt} \quad (2)$$

$$\rightarrow L = \mu_0 n N A = \mu_0 n^2 vol. \quad (3)$$

$L$  is a solenoid inductance, in unit of [Henri: H], and  $vol. = LA$  is the volume of the solenoid.

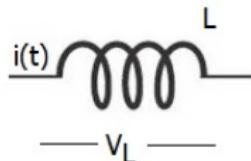


- Toroidal inductance ( on  $N$  turns, radius  $r$ , and cross section area  $A$ )

$$B(t) = \frac{\mu_0 N i(t)}{2\pi r} \rightarrow \Phi_B(t) = B(t)A = \mu_0 N A i(t) \quad (4)$$

$$\mathcal{E} = -N \frac{d\Phi_B}{dt} = -\frac{\mu_0 N^2 A}{2\pi r} \frac{di(t)}{dt} \equiv -V_L(t) \rightarrow L = \frac{\mu_0 N^2 A}{2\pi r}$$

- Inductor symbol



# Magnetic field energy

- ▶ The rate of electrical energy (power) dissipate into the solenoid inductor is

$$\mathcal{P} = iV_L = Li \frac{di}{dt} \equiv \frac{dU}{dt} \rightarrow U(t) = \frac{1}{2} Li^2 \quad (5)$$

- ▶ Rewrite in terms of the magnetic field strength  $B$  of the solenoid

$$\mathcal{P} = -i\mathcal{E} = Li \frac{di}{dt} = \frac{dU}{dt} \rightarrow U = \frac{1}{2} Li^2 \quad (6)$$

$$i = \frac{B}{\mu_0 n} \rightarrow U = \frac{1}{2} \frac{LB^2}{\mu_0^2 n^2} = \frac{1}{2\mu_0} B^2 \text{vol.} \quad (7)$$

$$u = \frac{U}{\text{vol.}} = \frac{1}{2\mu_0} B^2 \quad (8)$$

This is the energy density of the magnetic field strength  $B$

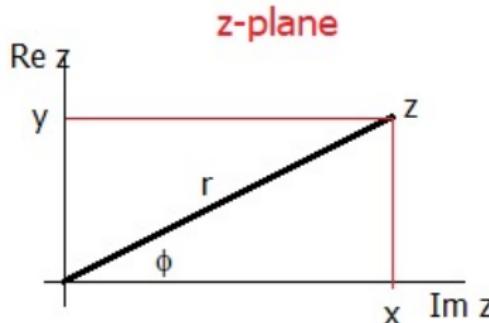
# Complex impedance

- ▶ complex number

$$z = x + iy \rightarrow \operatorname{Re}(z) = x, \operatorname{Im}(z) = y, i^2 = -1 \quad (9)$$

- ▶ Polar form

$$\begin{aligned} z &= re^{i\phi}, \quad r = \sqrt{x^2 + y^2}, \quad \phi = \tan^{-1} \frac{y}{x} \\ x &= r \cos \phi, \quad y = r \sin \phi \end{aligned} \quad (10)$$



► Complex impedance  $Z$

$$Z = R + iY, \quad R - \text{resistance}, \quad Y - \text{admittance} \quad (11)$$

► Ohm's law for AC current, with sinusoidal current

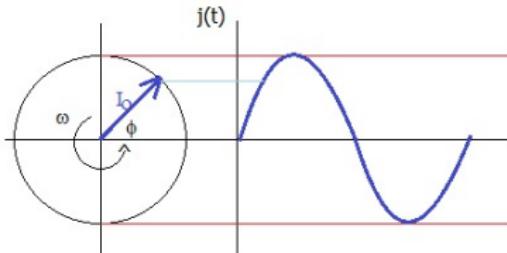
$$j(t) = I_0 \sin(\omega t) = \text{Im} (I_0 e^{i\omega t}) \quad v_Z = jZ \quad (12)$$

$$v_R = jZ_R = jR \quad \rightarrow Z_R = R \quad (13)$$

$$v_L = jZ_L = L \frac{dj}{dt} = i\omega L j \quad \rightarrow Z_L = i\omega L \quad (14)$$

$$v_C = \frac{1}{C} q(t) = \frac{1}{C} \int j(t) dt = \frac{1}{i\omega C} j \quad \rightarrow Z_C = \frac{1}{i\omega C} \quad (15)$$

► Phasor of AC current

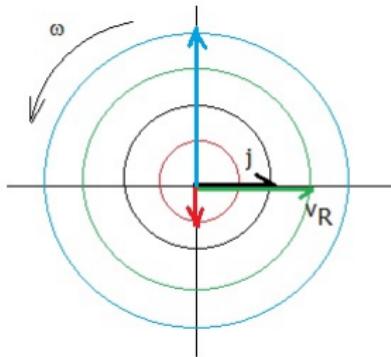


- Phasor of  $v_Z$ , with  $i = e^{i\pi/2}$  and  $-i = e^{-i\pi/2}$

$$v_R = jZ_R = I_0 R e^{i\omega t} = V_{0R} e^{i\omega t}, \quad V_{0R} = I_0 R \quad (16)$$

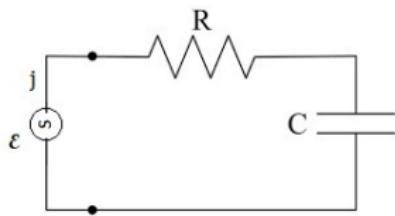
$$v_C = jZ_C = \frac{-i}{\omega C} I_0 e^{i\omega t} = V_{0C} e^{i\omega t - i\pi/2}, \quad V_{0C} = \frac{I_0}{\omega C} \quad (17)$$

$$v_L = jZ_L = i\omega L I_0 e^{i\omega t} = V_{0L} e^{i\omega t + i\pi/2}, \quad V_{0L} = \omega L I_0 \quad (18)$$



We observe that  $v_R$  is in-phase,  $v_C$  has lacking-phase, and  $v_L$  has leading-phase to the current  $j(t)$

- ▶ Let we have the current source  $j(t) = I_0 e^{i\omega t}$
- ▶ Let us determine the RC circuit



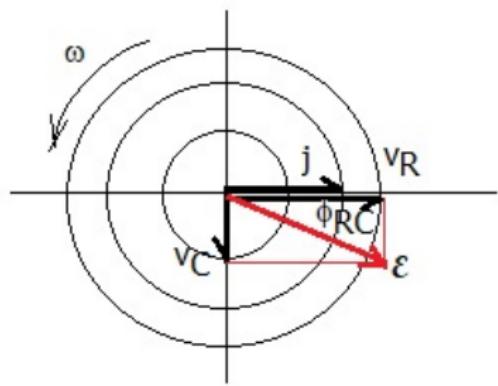
$$\mathcal{E} = v_R + v_C = (R - \frac{i}{\omega C})j$$

$$R - \frac{i}{\omega C} = \sqrt{R^2 + \frac{1}{(\omega C)^2}} e^{-i\phi_{RC}}$$

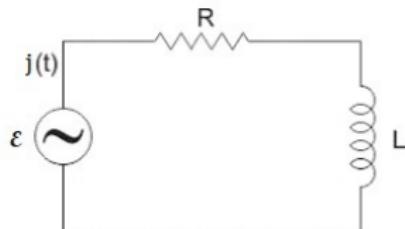
$$\phi_{RC} = \tan^{-1} \frac{1}{\omega C R}$$

$$\mathcal{E}(t) = \mathcal{E}_0 e^{i\omega t - i\phi_{RC}}$$

$$\mathcal{E}_0 = I_0 \sqrt{R^2 + \frac{1}{(\omega C)^2}}$$



- ▶ Let us determine the RL circuit



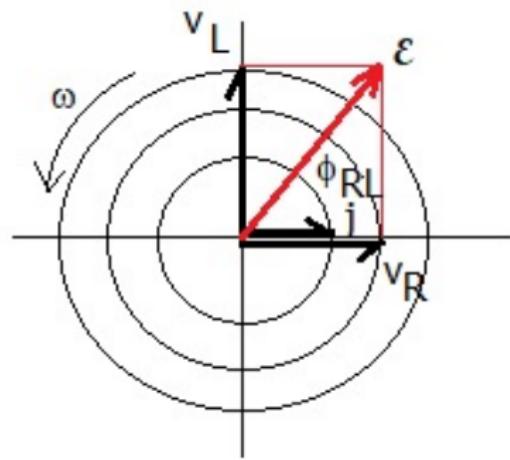
$$\mathcal{E} = v_R + v_L = (R + i\omega L)j$$

$$(R + i\omega L) = \sqrt{R^2 + (\omega L)^2} e^{i\phi_{RL}}$$

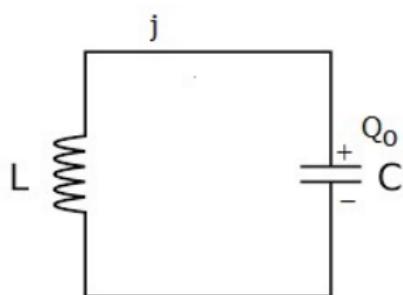
$$\phi_{RL} = \tan^{-1} \frac{\omega L}{R}$$

$$\mathcal{E} = \mathcal{E}_0 e^{i\omega t + i\phi_{RL}}$$

$$\mathcal{E}_0 = I_0 \sqrt{R^2 + \omega^2 L^2}$$



- ▶ Let us determine the LC circuit, start from a full stored capacitor



$$v_L + v_C = 0 \rightarrow i\omega Lj - \frac{i}{\omega C}j = 0$$

$$(\omega^2 - \frac{1}{LC})j = 0, \quad \omega = \frac{1}{\sqrt{LC}} \equiv \omega_0$$

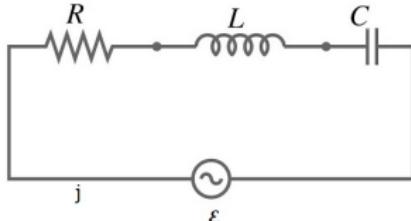
Simple oscillation of charge  $q(t)$  and the discharge current  $j(t) = -dq(t)/dt$

$$0 = v_C + v_L = \frac{1}{C}q(t) + L\frac{dj(t)}{dt} = \frac{1}{C}q(t) - L\frac{d^2q(t)}{dt^2}$$

$$\rightarrow \frac{d^2q}{dt^2} - \omega_0^2 q = 0 \rightarrow q(t) = Q_0 e^{i\omega_0 t}$$

## RLC resonance

- Let us determine the series RLC circuit



$$\begin{aligned}\mathcal{E} &= v_R + v_L + v_C \\ &= [R + i(\omega L - \frac{1}{\omega C})]j \\ &\equiv \mathcal{E}_0 e^{i\omega t + i\phi_{RLC}} \\ \mathcal{E}_0 &= I_0 Z_0\end{aligned}$$

where

$$Z_0 = \sqrt{R^2 + (\omega L - 1/\omega C)^2}, \text{ and } \phi_{RLC} = \tan^{-1} \left[ \frac{\omega L - 1/\omega C}{R} \right]$$

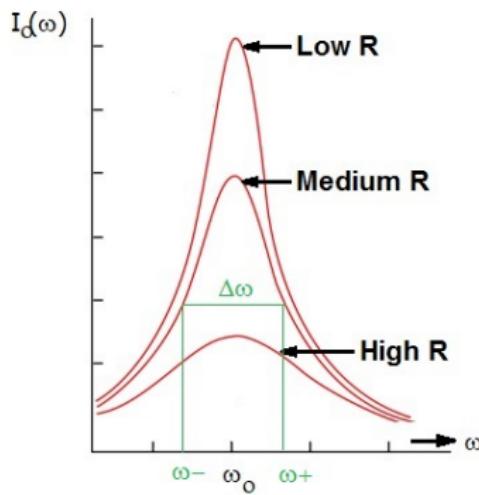
- Let  $\mathcal{E}_0$  is fixed, we observe that  $Z_0 = Z_0(\omega)$  and  $I_0 = I_0(\omega)$

$$I_0(\omega) = \frac{\mathcal{E}_0}{\sqrt{R^2 + (\omega L - 1/\omega C)^2}} \rightarrow I_{0,\max}(\omega_0) = \frac{\mathcal{E}_0}{R}$$

- We observe that there will be a maximum power deposit on  $R$ , with the mean power is

$$\bar{P}_{R,\max} = \frac{1}{2} I_{0,\max}^2 R = \frac{\mathcal{E}_0^2}{2R}$$

- ▶ This is a *resonance* condition, the resonance peak of the current  $I_0(\omega)$  is



- The *quality* of the peak is determined from the *full width half-maximum peak*  $\Delta\omega = \omega_+ - \omega_-$ , which is

$$\begin{aligned} \rightarrow \frac{\mathcal{E}_0^2}{4R} &= \frac{1}{2} \frac{\mathcal{E}_0^2 R}{R^2 + (\omega L - 1/\omega C)^2} \\ 2 &= 1 + \frac{(\omega L - 1/\omega C)^2}{R^2} \rightarrow \left( \frac{L}{R\omega} \right)^2 (\omega^2 - \omega_0^2)^2 = 1 \\ \rightarrow \omega_{\pm}^2 - \omega_0^2 &= \pm \frac{R\omega_{\pm}}{L} \end{aligned} \quad (19)$$

Solving for  $\omega_{\pm}$ , we get

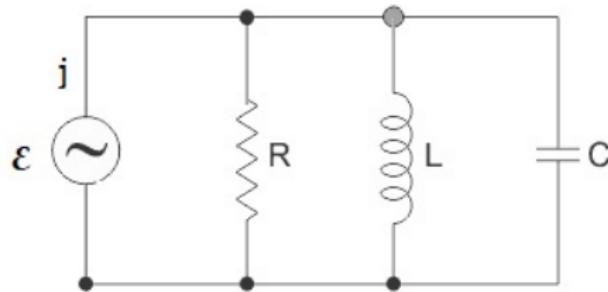
$$\omega_{\pm} = \pm \frac{R}{2L} + \sqrt{\left( \frac{R}{4L} \right)^2 + \omega_0^2} \quad (20)$$

$$\rightarrow \Delta\omega = \omega_+ - \omega_- = \frac{R}{L} \quad (21)$$

- The quality Q factor is defined to be

$$Q = \frac{\omega_0}{\Delta\omega} = \frac{\omega_0 L}{R}, \quad R_{low} \rightarrow high \ Q, R_{high} \rightarrow low \ Q$$

► Exercise: Resonance of parallel RLC circuit ( $\omega_0$ ,  $Q$ )



# Electromagnetic waves

## ► Electromagnetic equations

Gauss's law  $\oint \vec{E} \cdot d\vec{A} = \frac{q}{\epsilon_0}, \oint \vec{B} \cdot d\vec{A} = 0$  (22)

Ampere's law  $\oint \vec{B} \cdot d\vec{l} = \mu_0 I$  (23)

Faraday's law 
$$\begin{aligned} \mathcal{E} &= \oint_{\partial A} \vec{E} \cdot d\vec{l} = -\frac{d\Phi_B}{dt} \\ &= -\frac{d}{dt} \int_A \vec{B} \cdot d\vec{A} \end{aligned}$$
 (24)

# Electromagnetic waves

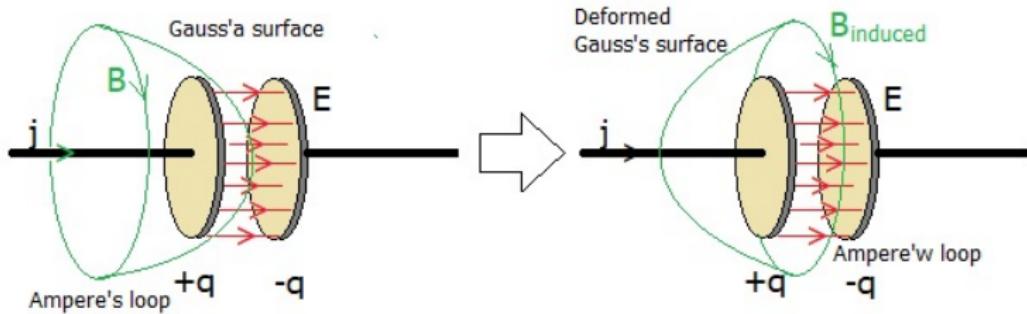
## ► Electromagnetic equations

$$\text{Gauss's law} \quad \oint \vec{E} \cdot d\vec{A} = \frac{q}{\epsilon_0}, \quad \oint \vec{B} \cdot d\vec{A} = 0 \quad (22)$$

$$\text{Ampere's law} \quad \oint \vec{B} \cdot d\vec{l} = \mu_0 I \quad (23)$$

$$\begin{aligned} \text{Faraday's law} \quad \mathcal{E} &= \oint_{\partial A} \vec{E} \cdot d\vec{l} = -\frac{d\Phi_B}{dt} \\ &= -\frac{d}{dt} \int_A \vec{B} \cdot d\vec{A} \end{aligned} \quad (24)$$

## ► Ampere-Maxwell's law



► Ampere-Maxwell's law (cont.)

$$\vec{B}_{induced} = \mu_0 \epsilon_0 \frac{d\Phi_E}{dt} = \mu_0 \epsilon_0 \frac{d}{dt} \int_A \vec{E} \cdot d\vec{A} \quad (25)$$

► Approved electromagnetic equations

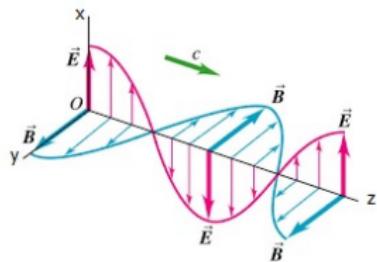
Gauss's law  $\oint \vec{E} \cdot d\vec{A} = \frac{q}{\epsilon_0}, \quad \oint \vec{B} \cdot d\vec{A} = 0 \quad (26)$

Faraday's law  $\oint_{\partial A} \vec{E} \cdot d\vec{l} = -\frac{d}{dt} \int_A \vec{B} \cdot d\vec{A} \quad (27)$

Ampere–  
–Maxwell's law  $\oint_{\partial A} \vec{B} \cdot d\vec{l} = \mu_0 i + \mu_0 \epsilon_0 \frac{d}{dt} \int_A \vec{E} \cdot d\vec{A} \quad (28)$

This set of equations is called *Maxwell's equation* in integral form.

# Plane electromagnetic wave

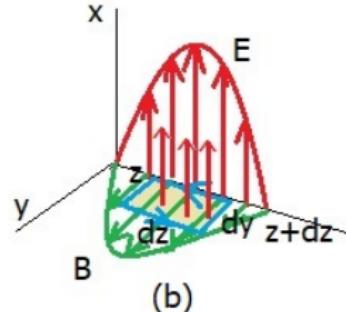
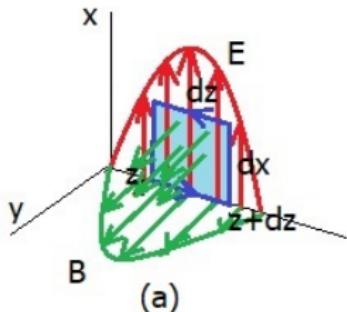


The wave functions are

$$\vec{E}(z, t) = \hat{x}E_0 \cos(kz - \omega t) \quad (29)$$

$$\vec{B}(z, t) = \hat{y}B_0 \cos(kz - \omega t) \quad (30)$$

To derive its wave equation, let us apply with Faraday's and Ampere-Maxwell's laws



We have

$$(a) : [E(z + dz, t) - E(z, t)]dx = -\frac{dB(z, t)}{dt}dxdz$$
$$\rightarrow \frac{dE(z, t)}{dz} = -\frac{dB(z, t)}{dt} \quad (31)$$

$$(b) : -[B(z + dz, t) - B(z, t)]dy = \mu_0 \epsilon_0 \frac{dE(z, t)}{dt} dy dz$$
$$\rightarrow \frac{dB(z, t)}{dz} = -\mu_0 \epsilon_0 \frac{dE(z, t)}{dt} \quad (32)$$

So that

$$\frac{d^2 E(z, t)}{dz^2} = -\frac{d}{dz} \frac{dB(z, t)}{dt} = -\frac{d}{dt} \frac{dB(z, t)}{dz} = \mu_0 \epsilon_0 \frac{d^2 E(z, t)}{dt^2}$$
$$\rightarrow \frac{d^2 E(z, t)}{dz^2} - \mu_0 \epsilon_0 \frac{d^2 E(z, t)}{dt^2} = 0 \quad (33)$$

We derive wave equation for electric field, with wave velocity

$$c_0^2 = \frac{1}{\mu_0 \epsilon_0} \rightarrow c_0 = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = 3 \times 10^8 m/s$$

Similar wave equation can be derived for magnetic field  $B(z, t)$ .

From (29,30), apply to the derived equations (31,32)

$$-kE_0 \sin(kz - \omega t) = -\omega B_0 \sin(kz - \omega t) \quad (34)$$

$$-kB_0 \sin(kz - \omega t) = -\frac{1}{c_0^2} \omega E_0 \sin(kz - \omega t) \quad (35)$$

$$\rightarrow c_0^2 \frac{B_0}{E_0} = \frac{\omega}{k} \rightarrow E_0 = c_0 B_0 \quad (36)$$

The amplitude of electric field is much larger than the amplitude of magnetic field inside the electromagnetic waves

## Electromagnetic energy-momentum

Energy of electric and magnetic fields in a volume  $V$  is

$$U = \frac{1}{2} \int \left( \frac{1}{\mu_0} B^2 + \epsilon_0 E^2 \right) dV = \epsilon_0 \int E^2 dV = \int u dV$$
$$\rightarrow u = \epsilon_0 E^2, \quad S = uc = \epsilon_0 c E^2 = \frac{1}{\mu_0} EB \quad (37)$$

$$\vec{S} = \frac{\vec{E} \times \vec{B}}{\mu_0}, \quad \text{its average : } \langle \vec{S} \rangle = \frac{\vec{E} \times \vec{B}}{2\mu_0} \quad (38)$$

This is called Poynting vector, in unit of  $[J/s \cdot m^2]$ .

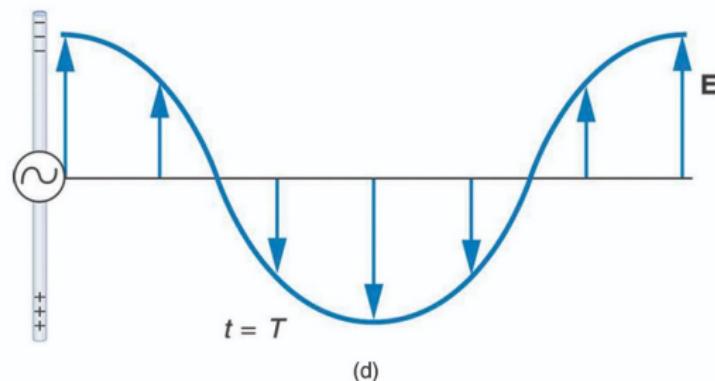
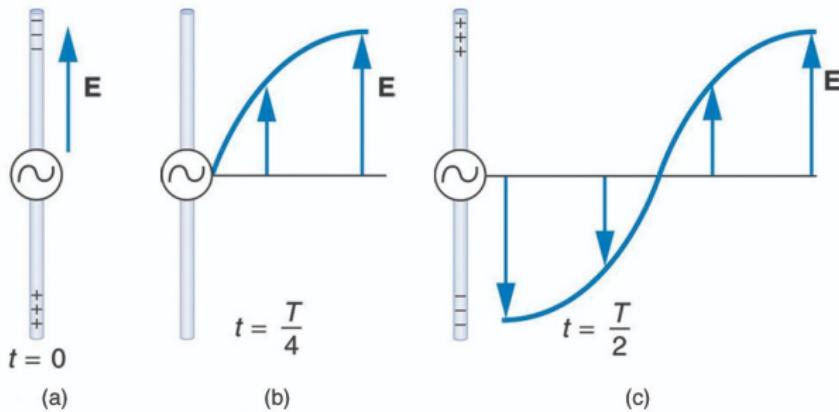
Radiation momentum and pressure are determined to be

$$p = \frac{U}{c} \rightarrow P = \frac{F}{A} = \frac{1}{A} \frac{dp}{dt} = \frac{1}{Ac} \frac{dU}{dt} = \frac{S}{c} \quad (39)$$

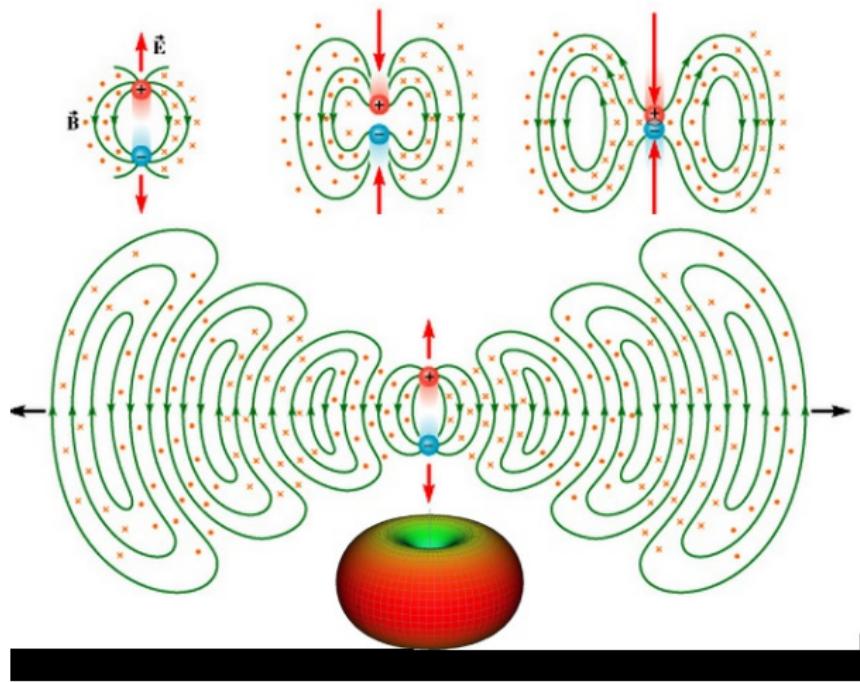
Momentum transfer for total absorption/reflection, will result to radiations pressure

$$dp_{absorp} = \frac{dU}{c} \rightarrow P = \frac{S}{c}, \quad dp_{reflect} = \frac{2dU}{c} \rightarrow P = \frac{2S}{c} \quad (40)$$

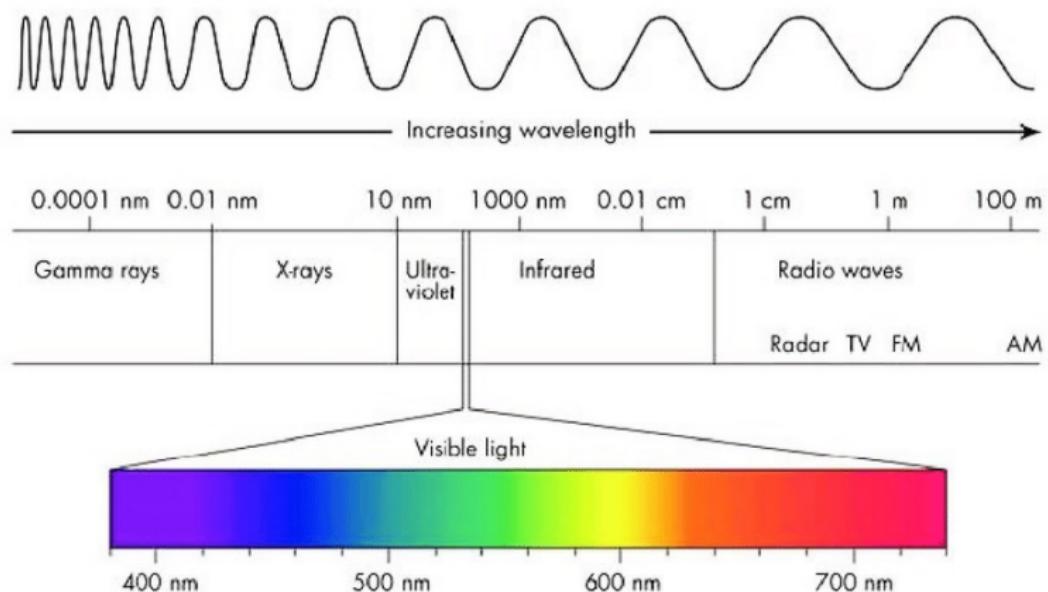
## ► Generation of EM waves - dipole antenna



## ► Dipole field in EM waves

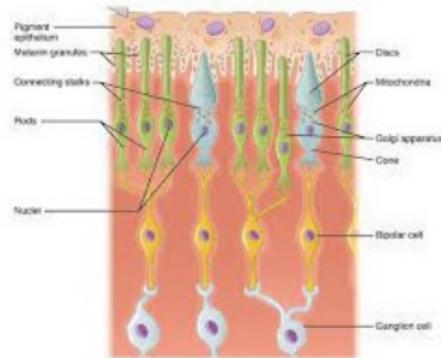
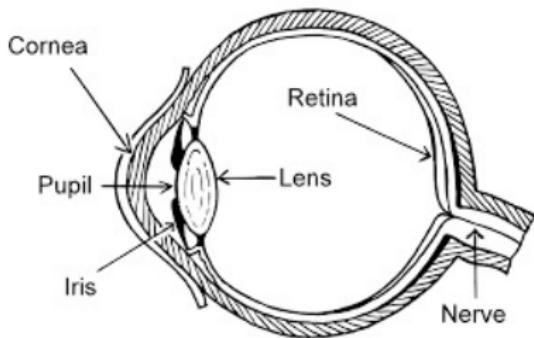


## ► Electromagnetic spectrum



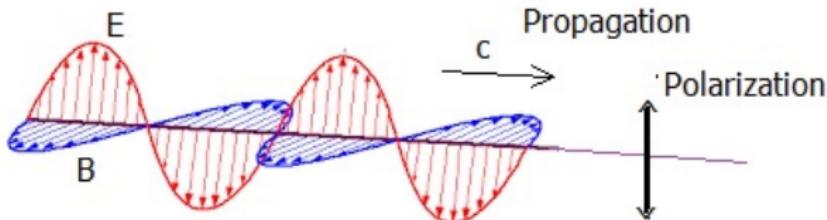
# Visible light

- Wavelength range of  $400\text{nm} - 70\text{nm}$
- Human eyes

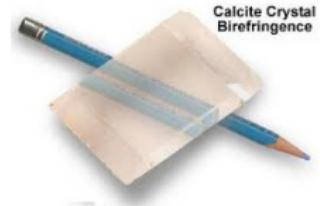
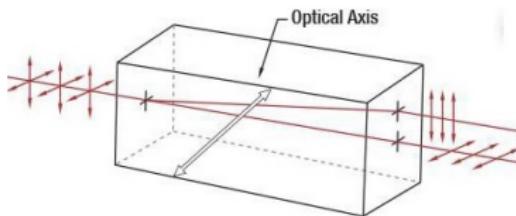


Rods	Cones
Coarse Detail	Fine Detail
Black and White	Colour
120 million cells	6 – 7 million cells
More sensitive to low light	Less sensitive to low light
Used for peripheral and night vision	Used for central and detailed vision

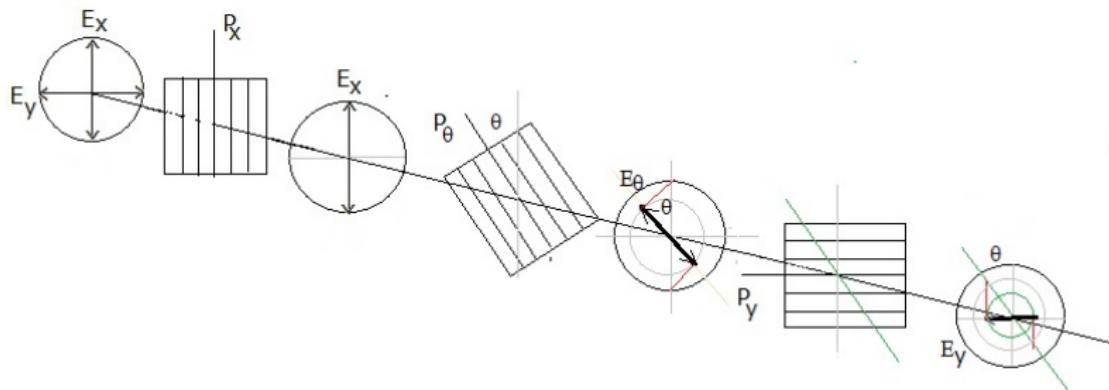
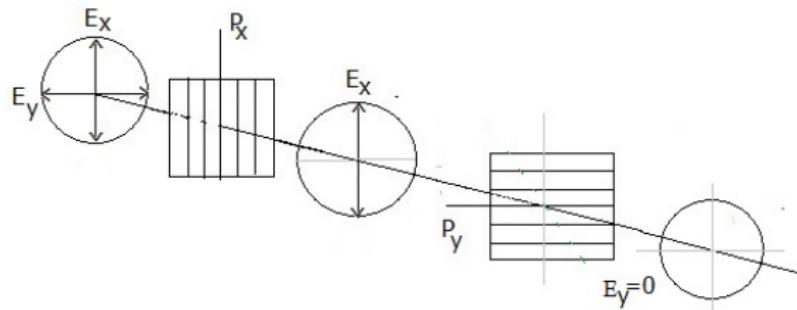
- ▶ Polarization - is defined in direction of electric field in EM waves



- ▶ Non-polarized light is light with polarization in all directions
- ▶ Production of polarized light
  - ▶ reflection at Brewster's angle
  - ▶ passing through birefringence crystal
  - ▶ passing through polaroid film

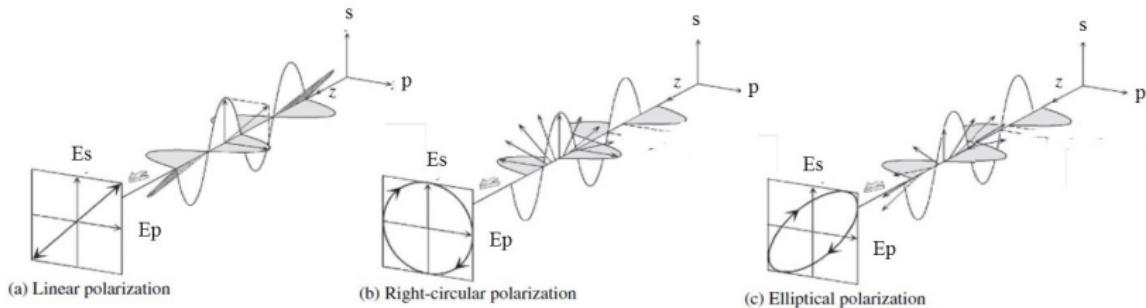


## ► Intrinsic polarization state of light



## ► Polarization states

- linear (x,y) polarization
- right/left circular polarization
- elliptical polarization



## ► Polaroid sun glasses

